

Stochastic Process

- Stochastic process (random process) is a time evolution of a statistical phenomenon according to probabilistic laws (function of time and not deterministic).
- We study a stochastic process at discrete and uniformly spaced instances of time.
- Discrete-time series (or time series) or discrete-time signal is one particular realization of a discrete-time stochastic process.
- A stochastic process is strictly stationary if its statistical properties are invariant over time, i.e., all the joint probability density functions of all times remain the same.

(1) Characterization of stochastic processes

- We consider a discrete-time stochastic process represented by the time series $u[n], u[n-1], \dots, u[n-M]$.
- The mean-value function is

$$\mu[n] = E[u[n]] = \sum_{u[n]=-\infty}^{\infty} u[n] f_{u[n]}$$

where E is the statistical expectation operator (ensemble average) and $f_{u[n]}$ is the probability mass function of $u[n]$.

- The autocorrelation function is

$$r[n, n-k] = E[u[n]u^*[n-k]], k = 0, \pm 1, \pm 2, \dots$$

where * denotes complex conjugation.

- The autocovariance function is

$$\begin{aligned} c[n, n-k] &= E[(u[n] - \mu[n])(u[n-k] - \mu[n-k])^*], k = 0, \pm 1, \pm 2, \dots \\ &= r[n, n-k] - \mu[n]\mu^*[n-k]. \end{aligned}$$

- For strictly stationary stochastic processes, all the joint probability density functions at all times remain the same. Therefore,

$$(1) \quad \forall n, \mu[n] = \mu, \quad r[n, n-k] = r[k], \quad c[n, n-k] = c[k].$$

$$(2) \quad r[0] = E[|u[n]|^2] \text{ is the } \underline{\text{mean-square value}} \text{ of } u[n].$$

$$(3) \quad c[0] = r[0] - \mu^2 = \sigma_u^2 \text{ is the variance of } u[n].$$

- In addition, if $\forall n, E[|u[n]|^2] < \infty$, the process is wide sense stationary (WSS) and the above three conditions hold.

(2) Mean ergodic theorem for WSS processes

- The expectation is "ensemble averages" of a stochastic process across the process.
- The "time average" is a long-term sample averages along the process. That is

$$\hat{\mu}[N] = \frac{1}{N} \sum_{n=0}^{N-1} u[n]$$

and $\hat{\mu}(N)$ itself is a random variable. We find that $\hat{\mu}(N)$ is unbiased since

$$E[\hat{\mu}[N]] = \mu, \forall N.$$

- We say the process $u[n]$ is mean ergodic in the mean-square error sense if

$$\lim_{N \rightarrow \infty} E[(\mu - \hat{\mu}[N])^2] = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{l=-N+1}^{N-1} \left(1 - \frac{|l|}{N}\right) c(l) = 0,$$

i.e., the process is asymptotically uncorrelated.

- The process is correlation ergodic in the mean-square sense, if

$$r[k, N] = \frac{1}{N} \sum_{n=0}^{N-1} u[n] u^*[n-k], 0 \leq k \leq N-1$$

and

$$\lim_{N \rightarrow \infty} E[(r[k] - \hat{r}[k, N])^2] = 0.$$

(3) Correlation matrix

- Let $M \times 1$ observation vector $\mathbf{u}[n]$ be

$$\mathbf{u}[n] = [u[n], u[n-1], \dots, u[n-M+1]]^T.$$

- We define the correlation matrix \mathbf{R} as

$$\begin{aligned} \mathbf{R} &= E[\mathbf{u}[n] \mathbf{u}^H[n]] \\ &= \begin{bmatrix} E[u[n]u^*[n]] & E[u[n]u^*[n-1]] & \cdots & E[u[n]u^*[n-M+1]] \\ E[u[n-1]u^*[n]] & E[u[n-1]u^*[n-1]] & \cdots & E[u[n-1]u^*[n-M+1]] \\ \vdots & \vdots & \vdots & \vdots \\ E[u[n-M+1]u^*[n]] & E[u[n-M+1]u^*[n-1]] & \cdots & E[u[n-M+1]u^*[n-M+1]] \end{bmatrix}. \end{aligned}$$

Then,

$$\mathbf{R} = \begin{bmatrix} r[0] & r[1] & \cdots & r[M-1] \\ r[-1] & r[0] & \cdots & r[M-2] \\ \vdots & \vdots & \vdots & \vdots \\ r[-M+1] & r[-M+2] & \cdots & r[0] \end{bmatrix}.$$

- The correlation matrix has the following properties.

(1) From the definition of \mathbf{R} , $\mathbf{R} = \mathbf{R}^H$. Therefore, $r[-k] = r^*[k]$. If $u[n]$ is real $\mathbf{R} = \mathbf{R}^T$. Therefore,

$$\mathbf{R} = \begin{bmatrix} r[0] & r[1] & \cdots & r[M-1] \\ r^*[1] & r[0] & \cdots & r[M-2] \\ \vdots & \vdots & \vdots & \vdots \\ r^*[M-1] & r^*[M-2] & \cdots & r[0] \end{bmatrix}.$$

(2) \mathbf{R} is Toeplitz iff $u[n]$ is WSS.

(3) \mathbf{R} is always nonnegative definite ($\forall \mathbf{x} \neq \mathbf{0}, \mathbf{x}^H \mathbf{R} \mathbf{x} \geq 0$) and almost always positive definite ($\forall \mathbf{x} \neq \mathbf{0}, \mathbf{x}^H \mathbf{R} \mathbf{x} > 0$). Therefore, \mathbf{R} is almost always nonsingular and $\det(\mathbf{R}) \neq 0$ and \mathbf{R}^{-1} exists.

(4) Let $\mathbf{u}^B[n] = [u[n-M+1], u[n-M+1], \dots, u[n]]^T$, then

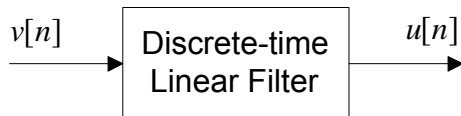
$$E[\mathbf{u}^B[n] \mathbf{u}^{BH}[n]] = \mathbf{R}^T.$$

(5) $\mathbf{R}_{M+1} = \begin{bmatrix} r[0] & \mathbf{r}^H \\ \mathbf{r} & \mathbf{R}_M \end{bmatrix} = \begin{bmatrix} \mathbf{R}_M & \mathbf{r}^{B*} \\ \mathbf{r}^{BT} & r[0] \end{bmatrix}$ with $\mathbf{r}^H = [r[1], r[2], \dots, r[M]]$ and

$$\mathbf{r}^{BT} = [r[-M], r[-M+1], \dots, r[-1]].$$

Stochastic Models

- Stochastic model is a representation of a stochastic process.
- We are interested in a linear process as a LTI system where

$$u[n] + \sum_{k=1}^M a_k^* u[n-k] = \sum_{k=0}^N b_k v[n-k]$$


with $v[n]$, white Gaussian noise as an input with $E[v[n]] = 0$ for all n , $r[0] = \sigma_v^2$, and $r[k] = 0$ for all $k \neq 0$.

(1) Autoregressive model (AR) model

- Time series $u[n], u[n-1], \dots, u[n-M]$ represents the realization of an AR of order M if

$$u[n] + \sum_{k=1}^M a_k^* u[n-k] = v[n] \quad \text{or} \quad u[n] = \sum_{k=1}^M a_k^* u[n-k] + v[n].$$

- With $a_0 = 1$, $\sum_{k=0}^M a_k^* u[n-k] = v[n]$, $H_{AR}(z) = \frac{U(z)}{V(z)} = \frac{1}{\sum_{k=0}^M a_k^* z^{-k}}$. Therefore, the filter is

an all-pole filter with infinite duration.

(2) Moving average (MA) model

- Time series $u[n], u[n-1], \dots, u[n-M]$ represents the realization of an MA of order N if

$$u[n] = \sum_{k=0}^N b_k v[n-k] \quad \text{with} \quad b_0 = 1.$$

- $H_{MA}(z) = \frac{U(z)}{V(z)} = \sum_{k=0}^N b_k z^{-k}$. Therefore, the filter is an all-zero filter with finite duration.

(3) Autoregressive-moving average (ARMA) model

- Time series $u[n], u[n-1], \dots, u[n-M]$ represents the realization of an ARMA of order (M, N) if

$$u[n] + \sum_{k=1}^M a_k^* u[n-k] = \sum_{k=0}^N b_k v[n-k] \quad \text{with} \quad b_0 = 1.$$

- $H_{ARMA}(z) = \frac{U(z)}{V(z)} = \frac{\sum_{k=0}^N b_k^* z^{-k}}{\sum_{k=0}^M a_k^* z^{-k}}$. Therefore, the filter is a mixed type with infinite duration.

(4) Wold decomposition

- Any stationary discrete-time stochastic process $x[n]$ can be expressed as

$$x[n] = u[n] + s[n]$$

where

(1) $u[n]$ and $s[n]$ are uncorrelated,

(2) $u[n]$ is a general linear process represented by the MA model:

$$u[n] = \sum_{k=0}^N b_k^* v[n-k] \quad \text{with } b_0 = 1, \text{ and } \sum_{k=0}^N |b_k|^2 < \infty$$

with white-noise process $v[n]$ such that $E[v[n]s^*[k]] = 0$ for all (n, k) ,

(3) $s[n]$ is a predictable process, that is, the process can be predicted from its own past with zero prediction variance.

(5) Yule-Walker equation (AR model)

- Unique description of an AR model of order M :

(1) The AR coefficients a_1, a_2, \dots, a_M and

(2) The variance σ_v^2 of $v[n]$.

- Consider the AR model $\sum_{k=0}^M a_k^* u[n-k] = v[n]$ with $a_0 = 1$. Multiply both sides by $u^*[n-l]$ and take expectations. Then,

$$E\left[\sum_{k=0}^M a_k^* u[n-k] u^*[n-l]\right] = E[v[n] u^*[n-l]], \text{ or}$$

$$\sum_{k=0}^M a_k^* r[l-k] = 0 \quad \text{for } l > 0.$$

For $l = 1, 2, \dots, M$, we have

$$\begin{aligned} l = 1 &\Rightarrow a_0^* r[1] + a_1^* r[0] + a_2^* r[-1] + \dots + a_M^* r[1-M] = 0 \\ l = 2 &\Rightarrow a_0^* r[2] + a_1^* r[1] + a_2^* r[0] + \dots + a_M^* r[2-M] = 0 \\ &\vdots \\ l = M &\Rightarrow a_0^* r[M] + a_1^* r[M-1] + a_2^* r[M-2] + \dots + a_M^* r[0] = 0 \end{aligned}$$

or

$$\begin{bmatrix} r[0] & r[1] & \cdots & r[M-1] \\ r^*[1] & r[0] & \cdots & r[M-2] \\ \vdots & \vdots & \ddots & \vdots \\ r^*[M-1] & r^*[M-2] & \cdots & r[0] \end{bmatrix} \begin{bmatrix} w_1 \\ w_2 \\ \vdots \\ w_M \end{bmatrix} = \begin{bmatrix} r^*[1] \\ r^*[2] \\ \vdots \\ r^*[M] \end{bmatrix}$$

with $w_k = -a_k$. In short,

$$\mathbf{R}\mathbf{w} = \mathbf{r} \quad \text{or} \quad \mathbf{w} = \mathbf{R}^{-1}\mathbf{r}.$$

- Note that $E[v[n]u^*[n]] = E[v[n]v^*[n]] = \sigma_v^2$ and $\sigma_v^2 = \sum_{k=0}^M a_k r[k]$.

Power Spectral Density

- Autocorrelation function: time-domain description of the 2nd order statistics
- Power spectral density (or power spectrum or spectrum): frequency-domain description of the 2nd order statistics

(1) Power spectral density (PSD)

- Consider a discrete-time time series of infinite duration;

$$\cdots u[n-M] \cdots u[0] \cdots u[n] u[n+1] \cdots$$

- For a segment of length N , define $u_N[n] = \begin{cases} u[n], & n = 0, 1, \dots, N-1 \\ 0, & \text{otherwise} \end{cases}$. Then,

$$U_N(\omega) = \sum_{n=0}^{N-1} u_N[n] e^{-j\omega n}, \quad U_N^*(\omega) = \sum_{k=0}^{N-1} u_N^*[k] e^{j\omega k}, \quad \text{and}$$

$$|U_N(\omega)|^2 = \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} u_N[n] u_N^*[k] e^{-j\omega(n-k)}.$$

- We take expected values of both sides as

$$E[|U_N(\omega)|^2] = \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} E[u_N[n] u_N^*[k]] e^{-j\omega(n-k)} = \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} r_N[n-k] e^{-j\omega(n-k)}.$$

$$\text{Since we have } r_N[n-k] = \begin{cases} E[u_N[n] u_N^*[k]] = r[n-k] & \text{for } 0 \leq (n, k) \leq N-1 \\ 0 & \text{otherwise} \end{cases},$$

$$E[|U_N(\omega)|^2] = \sum_{n=0}^{N-1} \sum_{k=0}^{N-1} r[n-k] e^{-j\omega(n-k)}.$$

$$\text{Let } l = n - k, \quad \frac{1}{N} E[|U_N(\omega)|^2] = \sum_{l=-N+1}^{N-1} \left(1 - \frac{|l|}{N}\right) r[l] e^{-j\omega l} = \sum_{l=-N+1}^{N-1} w_B[l] r[l] e^{-j\omega l} \quad \text{where}$$

$w_B[l]$ is the Barlett window. Note that $w_B[l] \rightarrow 1$ as $N \rightarrow \infty$.

- We define the periodogram of the windowed time series $u_N[n]$ as

$$P_N(\omega) = \frac{1}{N} |U_N(\omega)|^2.$$

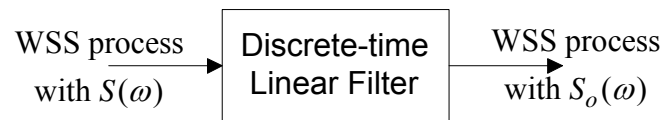
- We define the power spectral density of a WSS discrete-time stochastic process as

$$S(\omega) = \lim_{N \rightarrow \infty} E[P_N(\omega)] = \sum_{l=-\infty}^{\infty} r[l] e^{-j\omega l} \quad \text{with } r[l] = E[u[n] u^*[n-l]].$$

(2) Properties of PSD

- $S(\omega) = \sum_{l=-\infty}^{\infty} r[l]e^{-j\omega l}$, $-\pi < \omega \leq \pi \Leftrightarrow r[l] = \frac{1}{2\pi} \int_{-\pi}^{\pi} S(\omega)e^{j\omega l} d\omega$, $l = 0, \pm 1, \pm 2, \dots$
- $S(\omega + 2\pi k) = S(\omega)$ for integer k . $-\pi < \omega \leq \pi$ is the Nyquist interval.
- PSD of a discrete-time WSS process is real since $r[-k] = r^*[k]$.
- PSD of a real-valued discrete-time WSS process is even (i.e., symmetric), i.e., $S(\omega) = S(-\omega)$.
- $r[0] = \frac{1}{2\pi} \int_{-\pi}^{\pi} S(\omega)d\omega$: mean-square value or expected power across 1Ω resistor
- $S(\omega) \geq 0$ for all ω

(3) Transmission of a WSS process through a LTI system



- For a stable LTI system, $S_o(\omega) = |H(e^{j\omega})|^2 S(\omega)$.

Eigenanalysis of Correlation Matrix

- Consider an $M \times M$ Hermitian matrix \mathbf{R} which is a correlation matrix of a WSS process.

(1) Eigenvalue and eigenvector

- If $\mathbf{R}\mathbf{q} = \lambda\mathbf{q}$, λ is an *eigenvalue* and \mathbf{q} is the corresponding *eigenvector*. That is, \mathbf{q} is *invariant in direction* from the linear transformation by \mathbf{R} .
- The solution of the characteristic equation, $\det(\mathbf{R} - \lambda\mathbf{I}) = 0$ provide all λ_i and \mathbf{q}_i .

(2) Properties of eigenvalues and eigenvectors of \mathbf{R} from WSS process

- $\mathbf{R}^k \mathbf{q} = \lambda^k \mathbf{q}$
- $\{\lambda_i\}_{i=1}^M$ are all *real* and *nonnegative*.
- If $\{\lambda_i\}_{i=1}^M$ are distinct, $\{\mathbf{q}_i\}_{i=1}^M$ are linearly independent.
- If $\{\lambda_i\}_{i=1}^M$ are distinct, $\{\mathbf{q}_i\}_{i=1}^M$ are orthogonal. $\{\mathbf{q}_i\}_{i=1}^M$ is an orthogonal basis of $\text{span}(\mathbf{R})$.
- If $\{\lambda_i\}_{i=1}^M$ are distinct and $\{\mathbf{q}_i\}_{i=1}^M$ are normalized so that $\mathbf{q}_j^H \mathbf{q}_j = \delta_{ij}$,

$$\mathbf{Q}^H \mathbf{R} \mathbf{Q} = \mathbf{\Lambda} \quad (\text{unitary similarity transform})$$

with $\mathbf{Q} = [\mathbf{q}_1, \mathbf{q}_2, \dots, \mathbf{q}_M]$ and $\mathbf{\Lambda} = \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_M)$. Since $\mathbf{R}\mathbf{Q} = \mathbf{Q}\mathbf{\Lambda}$, $\mathbf{Q}^H \mathbf{Q} = \mathbf{I}$ or $\mathbf{Q}^{-1} = \mathbf{Q}^H$ (unitary matrix),

$$\mathbf{R} = \mathbf{Q}\mathbf{\Lambda}\mathbf{Q}^H = \sum_{i=1}^M \lambda_i \mathbf{q}_i \mathbf{q}_i^H \quad (\text{spectral theorem or Mercer's theorem}).$$

- $\text{tr}(\mathbf{Q}^H \mathbf{R} \mathbf{Q}) = \text{tr}(\mathbf{R} \mathbf{Q} \mathbf{Q}^H) = \text{tr}(\mathbf{R}) = \text{tr}(\mathbf{\Lambda}) = \sum_{i=1}^M \lambda_i$.
- The condition number of \mathbf{R} is $\chi(\mathbf{R}) = \|\mathbf{R}\| \|\mathbf{R}^{-1}\| = \frac{\lambda_{\max}}{\lambda_{\min}}$. \mathbf{R} is ill-conditioned if $\chi(\mathbf{R})$ is large.
- $S_{\min} \leq \lambda_i \leq S_{\max}$ where S_{\min} and S_{\max} are the minimum and maximum of the power

spectral density of the process. Therefore, $\chi(\mathbf{R}) = \frac{\lambda_{\max}}{\lambda_{\min}} \leq \frac{S_{\max}}{S_{\min}}$.

- If $\lambda_1 \leq \lambda_2 \leq \dots \leq \lambda_M$, $\lambda_k = \min_{\substack{\dim(T)=k \\ \mathbf{x} \in T \\ \mathbf{x} \neq 0}} \max \frac{\mathbf{x}^H \mathbf{R} \mathbf{x}}{\mathbf{x}^H \mathbf{x}}, k = 1, 2, \dots, M$.
- (Karhunen-Loeve expansion) $\mathbf{u}[n] = \sum_{i=1}^M c_i[n] \mathbf{q}_i$ with $c_i[n] = \mathbf{q}_i^H \mathbf{u}[n]$ for $i = 1, 2, \dots, M$.

Note

$$E[c_i[n]] = 0 \text{ for } i = 1, 2, \dots, M \text{ and } E[c_i[n]c_j^*[n]] = \begin{cases} \lambda_i, & i = j \\ 0, & i \neq j \end{cases}$$

Also,

$$\sum_{i=1}^n |c_i[n]|^2 = \|\mathbf{u}[n]\|^2 \text{ and } E[|c_i[n]|^2] = \lambda_i \text{ for } i = 1, 2, \dots, M.$$

(3) Low-rank modeling or subspace decomposition

- Assume $\lambda_1 > \lambda_2 > \dots > \lambda_M$ and $\{\lambda_i\}_{i=p+1}^M$ are all small, then we can approximate $\mathbf{u}[n]$ by

$$\hat{\mathbf{u}}[n] = \sum_{i=1}^p c_i[n] \mathbf{q}_i, p < M.$$

- $\text{span}\{\mathbf{q}_i\}_{i=1}^p$ is a feature space whereas $\text{span}\{\mathbf{q}_i\}_{i=1}^M$ is a data space.
- The reconstruction error vector is

$$\mathbf{e}[n] = \hat{\mathbf{u}}[n] - \mathbf{u}[n] = \sum_{i=p+1}^M c_i[n] \mathbf{q}_i.$$

- The mean-square error is

$$\varepsilon = E[\|\mathbf{e}[n]\|^2] = \sum_{i=p+1}^M \lambda_i.$$

(4) Eigenfilter

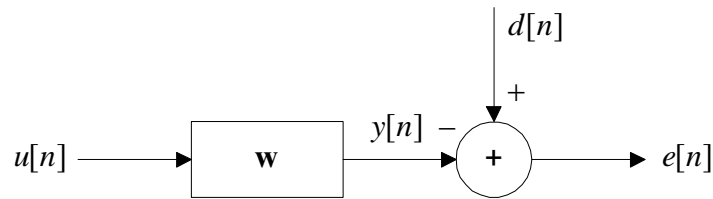
- We want to maximize the output signal-to-noise ratio (SNR).
- Eigendecomposition of the correlation matrix \mathbf{R} is performed.
- The eigenvector \mathbf{q}_{\max} which corresponds to the maximal eigenvalue defines the impulse response of the optimal filter.
- This eigenfilter maximizes the output SNR for a random signal.
- This corresponds to the matched filter for a deterministic signal.

Wiener Filter

Linear optimal filtering based on known statistics

$u[n]$ and $d[n]$ are jointly WSS

(1) Formulation and solution



Let $\mathbf{w} = [w_0 w_1 \cdots w_{M-1}]^T \in \mathbf{C}^M$ and $\mathbf{u}[n] = [u[n] u[n-1] \cdots u[n-M+1]]^T$. Let $y[n] = \mathbf{w}^H \mathbf{u}[n]$ and $e[n] = d[n] - y[n]$. Define

$$J(\mathbf{w}) = E\{|e[n]|^2\} = E\{|d[n] - \mathbf{w}^H \mathbf{u}[n]|^2\} \quad \text{and} \quad \mathbf{w}_o = \arg \min_{\mathbf{w} \in \mathbf{C}^M} J(\mathbf{w}).$$

From the orthogonality principle (OP),

$$E\{\mathbf{u}[n] e_o^*[n]\} = \mathbf{0} \quad \text{where} \quad e_o[n] = d[n] - \mathbf{w}_o^H \mathbf{u}[n].$$

Therefore, $E\{\mathbf{u}[n] (d[n] - \mathbf{w}_o^H \mathbf{u}[n])^H\} = E\{\mathbf{u}[n] d^*[n]\} - E\{\mathbf{u}[n] \mathbf{u}^H[n]\} \mathbf{w}_o = \mathbf{0}$ and

$$\mathbf{R} \mathbf{w}_o = \mathbf{p}$$

where $\mathbf{p} = E\{\mathbf{u}[n] d^*[n]\}$ and $\mathbf{R} = E\{\mathbf{u}[n] \mathbf{u}^H[n]\}$. This equation is called as the normal equation or Wiener-Hopf equation.

The solution of the normal equation is

$$\mathbf{w}_o = \mathbf{R}^{-1} \mathbf{p}$$

and

$$\hat{d}[n | U_n] = y_o[n] = \mathbf{w}_o^H \mathbf{u}[n].$$

Let $\sigma_d^2 = E\{|d[n]|^2\}$ and

$$\sigma_d^2 = E \left\{ \left| \hat{d}[n | U_n] \right|^2 \right\} = E \left\{ |y_o[n]|^2 \right\} = E \left\{ \mathbf{w}_o^H \mathbf{u}[n] \mathbf{u}^H[n] \mathbf{w}_o \right\} = \mathbf{w}_o^H \mathbf{R} \mathbf{w}_o = \mathbf{w}_o^H \mathbf{p} = \mathbf{p}^H \mathbf{w}_o$$

Since $d[n] = \hat{d}[n | U_n] + e_o[n]$ and $E \left\{ \hat{d}[n | U_n] e_o^*[n] \right\} = 0$ from OP,

$$J_{\min} = J(\mathbf{w}_o) = E \left\{ |e_o[n]|^2 \right\} = \sigma_d^2 - \sigma_d^2 \quad \text{and} \quad \varepsilon = \frac{J_{\min}}{\sigma_d^2} = 1 - \frac{\sigma_d^2}{\sigma_d^2}$$

(2) Error-performance surface in canonical coordinate

Let

$$\begin{aligned} J(\mathbf{w}) &= E \left\{ (d[n] - \mathbf{w}^H \mathbf{u}[n]) (d[n] - \mathbf{w}^H \mathbf{u}[n])^H \right\} \\ &= \sigma_d^2 - \mathbf{p}^H \mathbf{w} - \mathbf{w}^H \mathbf{p} + \mathbf{w}^H \mathbf{R} \mathbf{w} \\ &= \sigma_d^2 + (\mathbf{w} - \mathbf{w}_o)^H \mathbf{R} (\mathbf{w} - \mathbf{w}_o) - \mathbf{w}_o^H \mathbf{R} \mathbf{w}_o \\ &= \sigma_d^2 + (\mathbf{w} - \mathbf{w}_o)^H \mathbf{R} (\mathbf{w} - \mathbf{w}_o) - \mathbf{p}^H \mathbf{R}^{-1} \mathbf{p} \end{aligned}$$

then $J_{\min} = J(\mathbf{w}_o) = \sigma_d^2 - \mathbf{p}^H \mathbf{R}^{-1} \mathbf{p}$. Since $\mathbf{R} = \mathbf{Q} \mathbf{\Lambda} \mathbf{Q}^H$, $\mathbf{\Lambda} = \text{diag}[\lambda_1 \lambda_2 \cdots \lambda_M]$, and

$\mathbf{Q} = [\mathbf{q}_1 \mathbf{q}_2 \cdots \mathbf{q}_M]$ with $\mathbf{R} \mathbf{q}_i = \lambda_i \mathbf{q}_i$,

$$\begin{aligned} J(\mathbf{w}) &= J_{\min} + (\mathbf{w} - \mathbf{w}_o)^H \mathbf{R} (\mathbf{w} - \mathbf{w}_o) \\ &= J_{\min} + \mathbf{v}^H \mathbf{\Lambda} \mathbf{v} \\ &= J_{\min} + \sum_{k=1}^M \lambda_k |v_k|^2 \end{aligned}$$

with the principal axis $\mathbf{v} = \mathbf{Q}^H (\mathbf{w} - \mathbf{w}_o)$.

Linear Prediction

- We assume a discrete-time WSS process.
- Consider a discrete-time time series of infinite duration;

$$\cdots u[n-M] \cdots u[0] \cdots u[n] u[n+1] \cdots$$

(1) Forward prediction

- Let $U_{n-1} = \text{span} \{u[n-k]\}_{k=1}^M$. Also let the *predicted value* of $u[n]$ and the *forward prediction error* be

$$\hat{u}[n|U_{n-1}] = \sum_{k=1}^M w_{f,k}^* u[n-k] \quad \text{and} \quad f_M[n] = u[n] - \hat{u}[n|U_{n-1}] \quad \text{with the tap-weight vector}$$

$$\mathbf{w}_f = [w_{f,1}, w_{f,2}, \dots, w_{f,M}]^T = \arg \min_{\mathbf{w}_f \in \mathbf{C}^M} E[|f_M[n]|^2].$$

- Let $\mathbf{u}[n-1] = [u[n-1], u[n-2], \dots, u[n-M]]^T$ and $d[n] = u[n]$, then

$$\hat{u}[n|U_{n-1}] = \mathbf{w}_f^H \mathbf{u}[n-1]. \quad \text{Compare this with Wiener filter.}$$

- From the normal equation or Wiener-Hopf equation, $\mathbf{R} \mathbf{w}_f = \mathbf{p}$ with

$$\mathbf{R} = E\{\mathbf{u}[n-1]\mathbf{u}^H[n-1]\} = \begin{bmatrix} r[0] & r[1] & \cdots & r[M-1] \\ r[-1] & r[0] & \cdots & r[M-2] \\ \vdots & \vdots & \ddots & \vdots \\ r[-M+1] & r[-M+2] & \cdots & r[0] \end{bmatrix} \quad \text{and}$$

$$\mathbf{p} = E\{\mathbf{u}[n-1]d^*[n]\} = E\{\mathbf{u}[n-1]u^*[n]\} = \begin{bmatrix} r[-1] \\ r[-2] \\ \vdots \\ r[-M] \end{bmatrix} = \mathbf{r},$$

$$\mathbf{w}_f = \mathbf{R}^{-1} \mathbf{r}.$$

- The forward prediction error power is

$$P_M = \min E[|f_M[n]|^2] = \sigma_d^2 - \mathbf{p}^H \mathbf{w}_0 = r[0] - \mathbf{r}^H \mathbf{w}_f.$$

- Connection with AR process;

$$\text{FLP:} \quad f_M[n] = u[n] - \mathbf{w}_f^H \mathbf{u}[n-1] = u[n] - \sum_{k=1}^M w_{f,k}^* u[n-k]$$

AR: $u[n] = v[n] - \sum_{k=1}^M a_k^* u[n-k]$ where $v[n]$ is white Gaussian noise.

- In FLP, if $f_M[n]$ becomes white, the prediction using the order M is optimal.
- Augmented Wiener-Hopf equation for FLP;

$$\begin{bmatrix} r[0] & \mathbf{r}^H \\ \mathbf{r} & \mathbf{R} \end{bmatrix} \begin{bmatrix} 1 \\ -\mathbf{w}_f \end{bmatrix} = \begin{bmatrix} P_M \\ 0 \end{bmatrix}.$$

(2) Backward prediction

- Let $U_n = \text{span}\{u[n-k]\}_{k=0}^{M-1}$. Also let the predicted value of $u[n-M]$ and the backward prediction error be

$$\hat{u}[n-1|U_n] = \sum_{k=1}^M w_{b,k}^* u[n-k+1] \quad \text{and} \quad b_M[n] = u[n-M] - \hat{u}[n-M|U_n]$$

with the tap-weight vector $\mathbf{w}_b = [w_{b,1}, w_{b,2}, \dots, w_{b,M}]^T = \arg \min_{\mathbf{w}_b \in \mathbf{C}^M} E[|b_M[n]|^2]$.

- Let $\mathbf{u}[n] = [u[n], u[n-1], \dots, u[n-M+1]]^T$ and $d[n] = u[n-M]$, then

$\hat{u}[n-M|U_n] = \mathbf{w}_b^H \mathbf{u}[n]$. Compare this with Wiener filter.

- From the normal equation or Wiener-Hopf equation, $\mathbf{R}\mathbf{w}_b = \mathbf{p}$ with

$$\mathbf{R} = E\{\mathbf{u}[n]\mathbf{u}^H[n]\} = \begin{bmatrix} r[0] & r[1] & \dots & r[M-1] \\ r[-1] & r[0] & \dots & r[M-2] \\ \vdots & \vdots & \ddots & \vdots \\ r[-M+1] & r[-M+2] & \dots & r[0] \end{bmatrix} \quad \text{and}$$

$$\mathbf{p} = E\{\mathbf{u}[n]d^*[n]\} = E\{\mathbf{u}[n]u^*[n-M]\} = \begin{bmatrix} r[M] \\ r[M-1] \\ \vdots \\ r[1] \end{bmatrix} = \mathbf{r}^{B*},$$

$$\mathbf{w}_b = \mathbf{R}^{-1} \mathbf{r}^{B*}.$$

- The backward prediction error power is

$$P_M = \min E[|b_M[n]|^2] = \sigma_d^2 - \mathbf{p}^H \mathbf{w}_0 = r[0] - \mathbf{r}^{BT} \mathbf{w}_b.$$